# Digital Signal Processing Inferring Frequency Response from Poles and Zeros

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### Rational Transfer Function

Recall, if we have a system with

$$H(z) = \frac{b_0 + b_1 z^{-1} + \dots + b_M z^{-M}}{a_0 + a_1 z^{-1} + \dots + a_N z^{-N}}$$

and the ROC includes the unit circle, then we have

$$H(e^{j\omega}) = \frac{b_0 + b_1 e^{-j\omega} + \dots + b_M e^{-j\omega M}}{a_0 + a_1 e^{-j\omega} + \dots + a_N e^{-j\omega N}}$$
(1)

which can be factored to write

$$H(e^{j\omega}) = \left(\frac{b_0}{a_0}\right) \frac{(1 - c_1 e^{-j\omega}) \cdots (1 - c_M e^{-j\omega})}{(1 - d_1 e^{-j\omega}) \cdots (1 - d_N e^{-j\omega})}$$
(2)

The MATLAB commands roots and poly allow easy conversion between (1) and (2).

### Magnitude Response

We can write the magnitude response as

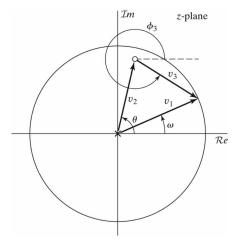
$$|H(e^{j\omega})| = \left| \frac{b_0}{a_0} \right| \frac{|1 - c_1 e^{-j\omega}| \cdots |1 - c_M e^{-j\omega}|}{|1 - d_1 e^{-j\omega}| \cdots |1 - d_N e^{-j\omega}|}$$
$$= \left| \frac{b_0}{a_0} \right| \frac{|e^{j\omega} - c_1| \cdots |e^{j\omega} - c_M|}{|e^{j\omega} - d_1| \cdots |e^{j\omega} - d_N|}$$

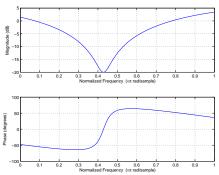
#### Intuition:

- $e^{j\omega}$  is a point on the unit circle.
- $ightharpoonup |e^{j\omega}-c_m|$  is the distance from the  $m^{\rm th}$  zero to that point on the unit circle.
- $ightharpoonup |e^{j\omega}-d_n|$  is the distance from the  $n^{\rm th}$  pole to that point on the unit circle.
- $ightharpoonup |H(e^{j\omega})|$  is large when  $e^{j\omega}$  is close to a pole and far from the zeros.
- $lackbox{ } |H(e^{j\omega})|$  is small when  $e^{j\omega}$  is close to a zero and far from the poles.

## 1st Order FIR Filter Example: $H(z) = 1 - az^{-1}$

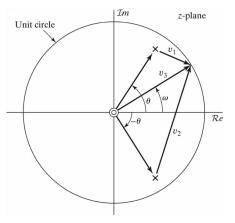
$$a = 0.9e^{j3\pi/7}$$

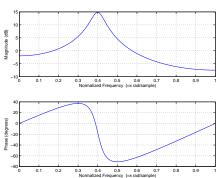




# 2nd Order IIR Filter Example: $H(z) = \frac{1}{1 - 2r\cos(\theta)z^{-1} + r^2z^{-2}}$

$$r=0.9$$
 and  $\theta=2\pi/5$ 





### Phase Response

We can write the phase response as

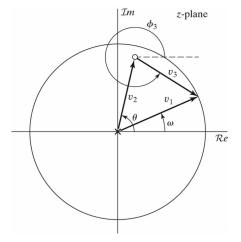
$$\angle H(e^{j\omega}) = \angle \frac{b_0}{a_0} + \sum_{m=1}^{M} \angle (1 - c_m e^{-j\omega}) - \sum_{n=1}^{N} \angle (1 - d_n e^{-j\omega})$$
$$= \angle \frac{b_0}{a_0} + \sum_{m=1}^{M} \left[ \angle (e^{j\omega} - c_m) - \angle e^{j\omega} \right] - \sum_{n=1}^{N} \left[ \angle (e^{j\omega} - d_n) - \angle e^{j\omega} \right]$$

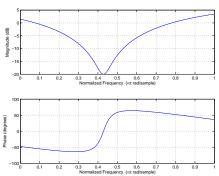
#### Intuition:

- $e^{j\omega}$  is a point on the unit circle.
- $ightharpoonup \angle(e^{j\omega}-c_m)$  is the angle of the vector from the  $m^{\rm th}$  zero to that point on the unit circle.
- $ightharpoonup \angle(e^{j\omega}-d_n)$  is the angle of the vector from the  $n^{\rm th}$  pole to that point on the unit circle.
- $\triangleright$   $\angle e^{j\omega} = \omega$ .

### 1st Order FIR Filter Example: $H(z) = 1 - az^{-1}$

$$a = 0.9e^{j3\pi/7}$$





# 2nd Order IIR Filter Example: $H(z) = \frac{1}{1 - 2r\cos(\theta)z^{-1} + r^2z^{-2}}$

$$r=0.9$$
 and  $\theta=2\pi/5$ 

