

# ECE531 Lecture 5: Detection of Discrete-Time Signals with Unknown Parameters

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# Introduction

- ▶ This lecture is about **parametric detection** problems for signals with **one or more unknown parameters**.
- ▶ We focus on the case of an  $N$ -sample discrete time observation  $Y \in \mathcal{Y}$  with binary hypotheses

$$\mathcal{H}_0 : Y \sim p_x(y) \text{ for } x \in \mathcal{X}_0$$

$$\mathcal{H}_1 : Y \sim p_x(y) \text{ for } x \in \mathcal{X}_1$$

- ▶ Our approach: Absorb the unknown parameters into our notion of the state  $x$  and the associated state space  $\mathcal{X}$ .
- ▶ In some textbooks, it is common to denote the random parameter(s) as  $\Theta$  and a realization of these parameters as  $\theta$ . We will stay consistent with our earlier notation, however, and use  $x$  here.

## Example

Suppose we have a known signal  $s \in \mathbb{R}^n$  and we have a communication system that transmits either  $s_0 = -s$  or  $s_1 = s$ . The signal arrives with unknown amplitude  $a > 0$  and is corrupted by zero-mean AWGN with variance  $\sigma^2$ . The hypotheses can be written as

$$\mathcal{H}_0 : Y \sim \mathcal{N}(as_0, \sigma^2 I)$$

$$\mathcal{H}_1 : Y \sim \mathcal{N}(as_1, \sigma^2 I)$$

for unknown  $a > 0$ . What is the state space  $\mathcal{X}$  and what are the sets  $\mathcal{X}_0$  and  $\mathcal{X}_1$ ?

An equivalent, but more clever way to write these hypotheses is

$$\mathcal{H}_0 : Y \sim \mathcal{N}(as, \sigma^2 I) \text{ with } a < 0$$

$$\mathcal{H}_1 : Y \sim \mathcal{N}(as, \sigma^2 I) \text{ with } a > 0$$

for unknown  $a \in \mathbb{R} \setminus \{0\}$ . What is the state space  $\mathcal{X}$  and what are the sets  $\mathcal{X}_0$  and  $\mathcal{X}_1$ ?

# Types of Detectors: What we Already Know

Problem: Find a decision rule to decide between

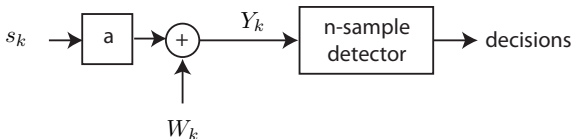
$$\mathcal{H}_0 : Y \sim \mathcal{N}(as, \sigma^2 I) \text{ with } a < 0$$

$$\mathcal{H}_1 : Y \sim \mathcal{N}(as, \sigma^2 I) \text{ with } a > 0$$

for unknown  $a \in \mathbb{R} \setminus 0$

- ▶ We know that these types of problems are **composite** binary hypothesis testing problems.
- ▶ Bayes decision rules involve computation of two discriminant functions (or commodity costs)  $g_0(y, \pi)$  and  $g_1(y, \pi)$  and subsequent comparison.
- ▶ Under the N-P criterion with significance level (or size)  $\alpha$ :
  - ▶ Our strategy: **reduce the composite HT problem to a simple one.**
  - ▶ Check for the existence of a UMP decision rule (check the critical region and/or monotone likelihood ratio) that maximizes  $P_{D,x}$  for all  $x \in \mathcal{X}_1$  subject to  $P_{\text{fp},x} \leq \alpha$  for all  $x \in \mathcal{X}_0$ .
  - ▶ If a UMP rule doesn't exist, we can often find an LMP decision rule.

# Example: Known Signal with Unknown Amplitude



- ▶ We would like to decide on the presence of a known signal  $s \in \mathbb{R}^n$  ( $\mathcal{H}_1$ ) versus the absence of the signal ( $\mathcal{H}_0$ ).
- ▶ The known signal  $s \in \mathbb{R}^n$  arrives at the detector corrupted by noise  $W \in \mathbb{R}^n$  and with unknown amplitude  $a \geq 0$ .
- ▶ We observe a realization  $y \in \mathbb{R}^n$  of the random variable  $Y = as + W$ .

$$a = 0 \text{ when no signal is present } (\mathcal{H}_0)$$

$$a > 0 \text{ when a signal is present } (\mathcal{H}_1)$$

- ▶ What kind of hypothesis testing problem is this? Binary, composite, one-sided, with a simple null hypothesis.
- ▶ What is the state space here?  $x = a \in \mathcal{X} = [0, \infty)$ .

# Bayes Detector for this Example

To develop a Bayes detector for this problem, we need 3 things:

1. We need a conditional pdf or pmf statistically describing the observations  $y \in \mathcal{Y}$  for each state  $x \in \mathcal{X}$ :  $p_x(y)$ .
2. We need a cost assignment for each state  $x \in \mathcal{X}$  and each hypothesis  $\mathcal{H}_i$ :  $C_0(x)$  and  $C_1(x)$ .
3. We need a prior (pdf) on the states:  $\pi(x)$ .

The Bayes risk of decision rule  $\rho$  in this case can be written as

$$\begin{aligned}
 r(\rho, \pi) &= \int_{\mathcal{X}} \pi(x) \int_{\mathcal{Y}} \rho^\top(y) C(x) p_x(y) dy dx \\
 &= \int_{\mathcal{Y}} \rho^\top(y) \underbrace{\left( \int_{\mathcal{X}} C(x) \pi(x) p_x(y) dx \right)}_{g(y, \pi) = [g_0(y, \pi), g_1(y, \pi)]^\top} dy
 \end{aligned}$$

We know how to solve this problem: For each  $y \in \mathcal{Y}$ , the Bayes decision rule just compares  $g_0(y, \pi)$  to  $g_1(y, \pi)$  and sets  $\delta^{B\pi}$  equal to the index of the smaller discriminant function, i.e. commodity cost.

# Bayes Detector for this Example

1. Suppose the noise is distributed as  $W \sim \mathcal{N}(0, \Sigma)$ . Conditioned on  $x = a$ , we can write the density of the vector observation as

$$p_{x=a}(y) = \frac{1}{(2\pi)^{n/2} |\Sigma|^{1/2}} \exp\left(-\frac{(y - as)^\top \Sigma^{-1} (y - as)}{2}\right)$$

2. Suppose also that we have the UCA.
3. Suppose finally that we have the prior

$$\pi(x) = \pi_0 \delta(x) + (1 - \pi_0) (u(x) - u(x - 1))$$

where  $u(x) = 1$  for all  $x > 0$  and  $u(x) = 0$  otherwise.

Note that the state space here is  $\mathcal{X} = [0, 1]$ .

In this sort of problem, it is convenient to transform the observations (and decorrelate the noise) using our coordinate transformation trick.

```

% Bayes example plots
%-----
% User variables
%-----
pi0 = 0.75;           % prior state parameter
ybartest = -1:0.1:2; % space of observations to test (in transformed coordinate space)
sbar = [1.5;0.5];    % signal vector (in transformed coordinate space)
%-----

v = sbar'*sbar/2;
syms a real
discrimratio = zeros(length(ybartest),length(ybartest));

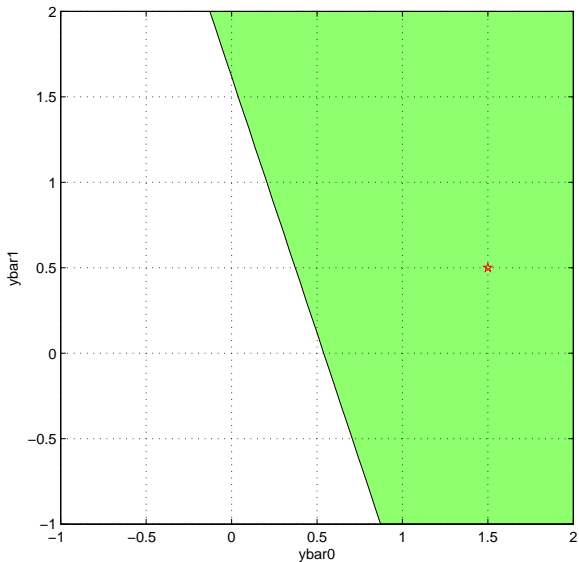
i0 = 0;
for y0 = ybartest
    i0 = i0+1;
    i1 = 0;
    for y1 = ybartest
        i1 = i1+1;
        ybar = [y0;y1];
        u = sbar'*ybar;
        f=exp(a*u-a^2*v);
        discrimratio(i1,i0) = (1-pi0)/pi0 * int(f,0,1);
    end
end

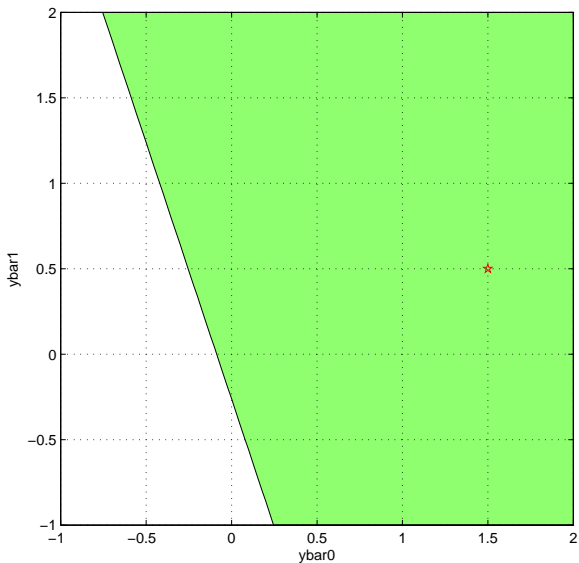
contourf(ybartest,ybartest,discrimratio,[1 1]);
xlabel('ybar0'); ylabel('ybar1');
hold on; plot(sbar(1),sbar(2),'rp'); hold off; grid on; axis square

```

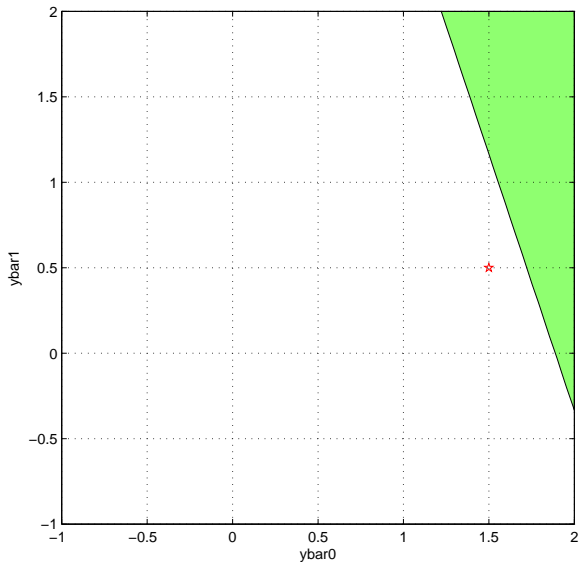


Bayes Decision Rule:  $\pi_0 = 0.5$  and  $\bar{s} = [1.5, 0.5]^\top$



Bayes Decision Rule:  $\pi_0 = 0.4$  and  $\bar{s} = [1.5, 0.5]^\top$ 

Bayes Decision Rule:  $\pi_0 = 0.75$  and  $\bar{s} = [1.5, 0.5]^T$



# Neyman-Pearson Detector

To develop a N-P detector for this problem, we need 2 things:

1. We need a conditional pdf or pmf statistically describing the observations  $y \in \mathcal{Y}$  for each state  $x \in \mathcal{X}$ :  $p_x(y)$ .
2. We need a significance-level for the test:  $P_{\text{fp}} \leq \alpha$ .

How should we approach the problem?

- ▶ Check to see if a UMP detector exists.
- ▶ If not, derive an LMP detector with good performance for  $\alpha$  near zero.
- ▶ A new option (“hybrid”): If we can specify a prior on the states, denoted as  $\pi(x)$ , we can reduce each composite hypothesis to a simple one by taking a weighted average of the density with respect to the prior, i.e.

$$\mathcal{H}_j' : Y \sim p_j(y) = \int_{\mathcal{X}_j} p_x(y) \pi_j(x) dx$$

where  $\pi_j(x) = \frac{\pi(x)}{\text{Prob}(x \in \mathcal{X}_j)}$  when  $x \in \mathcal{X}_j$  and is equal to zero otherwise.

Then testing  $\mathcal{H}_0'$  versus  $\mathcal{H}_1'$  is a **simple hypothesis testing problem** for which the N-P lemma gives a unique optimal solution.

# N-P Detector: Known Signal with Unknown Amplitude

Suppose the noise is distributed as  $W \sim \mathcal{N}(0, \sigma^2 I)$ . Conditioned on  $x = a$ , we can write the density of the vector observation as

$$\begin{aligned}
 p_{x=a}(y) &= \frac{1}{(2\pi\sigma^2)^{n/2}} \exp\left(-\frac{(y - as)^\top (y - as)}{2\sigma^2}\right) \\
 &= \prod_{k=0}^{n-1} p_{x=a}(y_k) \\
 &= \prod_{k=0}^{n-1} \frac{1}{\sqrt{2\pi}\sigma} \exp\left(-\frac{(y_k - as_k)^2}{2\sigma^2}\right)
 \end{aligned}$$

Note that any results we derive for this case can also be applied to the case when  $W \sim \mathcal{N}(0, \Sigma)$  by using our coordinate-transformation (decorrelation) trick.

# N-P Detector: Known Signal with Unknown Amplitude

To check for the existence of a UMP decision rule, let's set up a simple HT test:  $\mathcal{H}_0 : a = 0$  versus  $\mathcal{H}_1 : a = a_1 > 0$ .

The N-P lemma says that, for this simple HT problem, we can find an  $\alpha$ -level decision rule of the form

$$\rho(y) = \begin{cases} 1 & \ln(L(y)) > \ln v \\ \gamma & \ln(L(y)) = \ln v \\ 0 & \ln(L(y)) < \ln v \end{cases}$$

where  $\ln(L(y)) := \ln\left(\frac{p_1(y)}{p_0(y)}\right)$  with  $v \geq 0$  and  $\gamma \in [0, 1]$  chosen such that  $P_{\text{fp}} = \alpha$ .

Hence we need to compute the log-likelihood ratio...

# N-P Detector: Known Signal with Unknown Amplitude

The log-likelihood ratio for this simple HT test:

$$\begin{aligned}
 \ln(L(y)) &= \ln\left(\frac{p_1(y)}{p_0(y)}\right) \\
 &= \ln\left(\frac{\prod_{k=0}^{n-1} \exp\left(-\frac{(y_k - a_1 s_k)^2}{2\sigma^2}\right)}{\prod_{k=0}^{n-1} \exp\left(-\frac{y_k^2}{2\sigma^2}\right)}\right) \\
 &= \ln\left(\prod_{k=0}^{n-1} \exp\left(\frac{2y_k a_1 s_k - a_1^2 s_k^2}{2\sigma^2}\right)\right) \\
 &= \sum_{k=0}^{n-1} \frac{a_1}{\sigma^2} \left(y_k s_k - \frac{a_1 s_k^2}{2}\right) \\
 &= \frac{a_1}{\sigma^2} \left(s^\top y - \frac{a_1 \|s\|^2}{2}\right)
 \end{aligned}$$

Now we need to compute the threshold to achieve the false positive

# N-P Detector: Known Signal with Unknown Amplitude

The false-positive probability can be calculated as

$$\begin{aligned} P_{\text{fp}} &= \text{Prob}(\ln(L(Y)) > \ln v; a = 0) + \gamma \text{Prob}(\ln(L(Y)) = \ln v; a = 0) \\ &= \text{Prob}\left(s^\top Y > \frac{\sigma^2}{a_1} \ln v + \frac{a_1 \|s\|^2}{2}; a = 0\right) \end{aligned}$$

How is  $s^\top Y$  distributed when  $a = 0$ ?

Hence,  $v$  must be selected such that

$$P_{\text{fp}} = Q\left(\frac{\frac{\sigma^2}{a_1} \ln v + \frac{a_1 \|s\|^2}{2}}{\|s\| \sigma}\right) = \alpha$$

This implies the a value for  $v$ .



# N-P Detector: Known Signal with Unknown Amplitude

Does a UMP decision rule exist? To answer this question, we need to determine if the critical region  $\Gamma_1 = \{y \in \mathcal{Y} : \rho(y) \text{ decides } \mathcal{H}_1\}$  depends on our choice of  $a_1$ .

We decide  $\mathcal{H}_1$  for sure when

$$s^\top y > \frac{\sigma^2}{a_1} \ln v + \frac{a_1 \|s\|^2}{2}$$

and we decide  $\mathcal{H}_1$  with probability  $\gamma$  when

$$s^\top y = \frac{\sigma^2}{a_1} \ln v + \frac{a_1 \|s\|^2}{2}$$

(which happens with probability zero).

Critical region depends on  $a_1 \Rightarrow$  no UMP decision rule. ☹

What should we do now?

# N-P Detector: Known Signal with Unknown Amplitude

Hold on. Does the critical region  $\Gamma_1 = \{y \in \mathbb{R}^n : s^\top y > \frac{\sigma^2}{a_1} \ln v + \frac{a_1 \|s\|^2}{2}\}$  really depend on  $a_1$ ?

Given  $0 \leq \alpha \leq 1$  and  $t > 0$ , the unique solution to  $Q(z/t) = \alpha$  is  $z = tQ^{-1}(\alpha)$ . Hence, the unique solution to

$$Q\left(\frac{\frac{\sigma^2}{a_1} \ln v + \frac{a_1 \|s\|^2}{2}}{\|s\| \sigma}\right) = \alpha$$

is

$$\frac{\sigma^2}{a_1} \ln v + \frac{a_1 \|s\|^2}{2} = \|s\| \sigma Q^{-1}(\alpha).$$

Hence the critical region for a significance-level  $\alpha$  N-P decision rule can be written as

$$\Gamma_1 = \{y \in \mathbb{R}^n : s^\top y > \|s\| \sigma Q^{-1}(\alpha)\}$$

Does this depend on  $a_1$ ? Does a UMP decision rule exist?

# N-P Detector: Known Signal with Unknown Amplitude

Here is our UMP decision rule:

$$\rho^{\text{UMP}}(y) = \begin{cases} 1 & s^\top y > \|s\| \sigma Q^{-1}(\alpha) \\ \gamma & s^\top y = \|s\| \sigma Q^{-1}(\alpha) \\ 0 & s^\top y < \|s\| \sigma Q^{-1}(\alpha) \end{cases}$$

How would this change if the noise was distributed as  $W \sim \mathcal{N}(0, \Sigma)$ ?

# N-P Detector: Known Signal with Unknown Amplitude

If the UMP detector did not exist or was too complicated, we could find an LMP detector for this example by comparing

$$\frac{d}{da} L_a(y) \Big|_{a=0}$$

to a threshold. When the noise samples are i.i.d., the likelihood ratio can be written as

$$\begin{aligned} L_a(y) &= \frac{p_{x=a}(y)}{p_{x=0}(y)} \\ &= \prod_{k=0}^{n-1} \frac{q(y_k - as_k)}{q(y_k)} \end{aligned}$$

where  $q(x)$  is the pmf/pdf of the  $k$ th noise sample. In our example,  $q(x) = \frac{1}{\sqrt{2\pi}\sigma} e^{-x^2/\sigma^2}$ . We'll continue our analysis here for i.i.d. noise with general distribution  $q(x)$ ...

# Neyman-Pearson Detector: LMP Decision Rule

Taking the derivative of  $L_a(y)$  with respect to  $a$  yields

$$\frac{d}{da} L_a(y) = - \sum_{k=0}^{n-1} s_k \frac{q'(y_k - a s_k)}{q(y_k)} \prod_{j \neq k} \frac{q(y_j - a s_j)}{q(y_j)}$$

where  $q'(x) = \frac{d}{dx} q(x)$ . Setting  $a = 0$  yields

$$\begin{aligned} \frac{d}{da} L_a(y)|_{a=0} &= - \sum_{k=0}^{n-1} s_k \frac{q'(y_k)}{q(y_k)} \prod_{j \neq k} \frac{q(y_j)}{q(y_j)} \\ &= - \sum_{k=0}^{n-1} s_k \frac{q'(y_k)}{q(y_k)} \\ &= \sum_{k=0}^{n-1} s_k h^{\text{LMP}}(y_k) \end{aligned}$$

# Neyman-Pearson Detector: LMP Decision Rule

The locally most powerful decision rule then takes the form

$$\rho^{\text{LMP}}(y) = \begin{cases} 1 & \sum_{k=0}^{n-1} s_k h^{\text{LMP}}(y_k) > \tau \\ \gamma & \sum_{k=0}^{n-1} s_k h^{\text{LMP}}(y_k) = \tau \\ 0 & \sum_{k=0}^{n-1} s_k h^{\text{LMP}}(y_k) < \tau \end{cases}$$

where  $\gamma$  and  $\tau$  are selected such that  $P_{\text{fp}} = \alpha$ .

In our Gaussian example,  $h^{\text{LMP}}(y_k) := -\frac{q'(y_k)}{q(y_k)} = \frac{2y_k}{\sigma^2}$ . Hence, if  $\tau' = \tau\sigma^2/2$ , the LMP decision rule then takes the form

$$\rho^{\text{LMP}}(y) = \begin{cases} 1 & s^\top y > \tau' \\ \gamma & s^\top y = \tau' \\ 0 & s^\top y < \tau' \end{cases}$$

with  $\tau'$  and  $\gamma$  selected so that  $P_{\text{fp}} = \alpha$ . How does this compare to the UMP decision rule?  $\rho^{\text{LMP}}(y) = \rho^{\text{UMP}}(y)$  when the noise is i.i.d. Gaussian.

# LMP Decision Rule for i.i.d. Laplacian Noise

For  $b > 0$ , the Laplacian density is given as  $q(x) = \frac{b}{2}e^{-b|x|}$ .

For Laplacian noise,

$$h^{\text{LMP}}(y_k) = -\frac{q'(y_k)}{q(y_k)} = b \operatorname{sgn}(x).$$

Hence, the locally most powerful decision rule can be written as

$$\rho^{\text{LMP}}(y) = \begin{cases} 1 & \sum_{k=0}^{n-1} s_k \operatorname{sgn}(y_k) > \tau \\ \gamma & \sum_{k=0}^{n-1} s_k \operatorname{sgn}(y_k) = \tau \\ 0 & \sum_{k=0}^{n-1} s_k \operatorname{sgn}(y_k) < \tau \end{cases}$$

In this case,  $\rho^{\text{LMP}}(y) \neq \rho^{\text{UMP}}(y)$ .

# Hybrid Approach: Average Density Over a Prior

An sub-optimal approach used in some textbooks is to reduce the composite HT problem to a simple HT problem by using the a prior  $\pi(x)$  to compute a single density as a weighted average of the family of densities associated with  $\mathcal{H}_j$ , i.e.

$$p_j(y) = \int_{\mathcal{X}_j} p_x(y)\pi_j(x) dx$$

where  $\pi_j(x) = \frac{\pi(x)}{\text{Prob}(x \in \mathcal{X}_j)}$  when  $x \in \mathcal{X}_j$  and is equal to zero otherwise. In this case, given a significance level  $\alpha$ , the N-P lemma tells us that a unique optimal solution must exist based on a threshold test of the likelihood ratio

$$\begin{aligned} L(y) &= \frac{p_1(y)}{p_0(y)} \\ &= \frac{\int_{\mathcal{X}_1} p_x(y)\pi_1(x) dx}{\int_{\mathcal{X}_0} p_x(y)\pi_0(x) dx}. \end{aligned}$$



# Remarks on the Hybrid Approach

- ▶ If there is some uncertainty as to the prior, the usual approach is to choose  $\pi(x)$  such that it gives as little information about  $x$  as possible, i.e. such that the simple HT problem has “maximum difficulty”.
- ▶ Suppose the unknown parameter is the phase of the received signal. What prior on this phase would give the least information?

$$\pi(x) \sim \mathcal{U}(0, 2\pi) \text{ or } \pi(x) \sim \mathcal{U}(-\pi, \pi)$$

This is commonly used in practice.

# Conclusions

- ▶ Detection of known signals in noise: **simple hypothesis testing**.
- ▶ Detection of signals with one or more unknown parameters in noise: **composite hypothesis testing**.
- ▶ Your textbook has lots of examples in Chapter 7. Unfortunately, most of them are solved with the GLRT.
- ▶ The GLRT is not covered in ECE531, but you should know that it is a suboptimal technique in which
  - ▶ The unknown parameters are first **estimated** (to be covered in the second half of this course)
  - ▶ The estimates are plugged in to the signal model so now all of the parameters are “known”
  - ▶ Simple hypothesis testing is performed

# Midterm Exam: What You Need to Know

- ▶ A basic understanding of the different types of hypothesis testing problems (binary,  $M$ -ary, simple, composite).
- ▶ Mathematical model of hypothesis testing problems.
- ▶ Intuition about “good” and “bad” decision rules.
- ▶ Lecture notes from lectures 1-4 (not including LMP detectors)
  - ▶ Bayesian hypothesis testing (binary and  $M$ -ary, simple and composite)
  - ▶ Neyman-Pearson hypothesis testing (binary, simple and composite)
  - ▶ Decorrelation of signals observed in correlated Gaussian noise.
  - ▶ Detection of known discrete-time signals (binary and  $M$ -ary)
  - ▶ Detection of discrete-time signals with random parameters (binary)
- ▶ Kay II Chapters 1, 3, 4, 6 but not including
  - ▶ GLRT, Rao, Wald detectors for composite HT problems
  - ▶ LMP decision rules
  - ▶ Anything having to do with estimation